



Today's Electrically Small Antennas: Over 100 Years of Development

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Small Antenna Background

The performance of any ESA is characterized and understood through the following:

- ❑ Impedance Match and Matching Technique
- ❑ Radiation Efficiency
- ❑ Directivity (Radiation Pattern Shape)
- ❑ Quality Factor - Q (bandwidth)

Impedance Matching

- ❑ Any electrically small antenna can be matched at any single frequency
 - ❑ Short vertical radiator
 - ❑ Horizontal radiator very close to metal ground
- ❑ Trade-offs with decreasing size
 - ❑ Bandwidth
 - ❑ Radiation efficiency
- ❑ Generally, the first objective is to tune the antenna's reactance as this facilitates easier and more effective matching
- ❑ Generally, the second objective is to match or transform the antenna's small resistance to the system characteristic resistance (typically 50Ω)

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Impedance Matching

- ❑ Oftentimes, we try to impedance match with the greatest radiation efficiency – not always a practical option
 - ❑ Repeatability issues in manufacturing
 - ❑ Detuning effects in surrounding environment
 - ❑ Bandwidth may not meet system requirements
- ❑ Many small antenna designs successfully achieve wider bandwidths with decreased efficiency – typically with lower gain at the lower operating frequencies

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Tuning – Achieving Self Resonance

Inductive Loading

Lumped element

Distributed (e.g. - increasing conductor length)

Capacitive Loading

Lumped element

Distributed (e.g. - top-hat, disk loading)

Combined Inductive/Inductive Loading

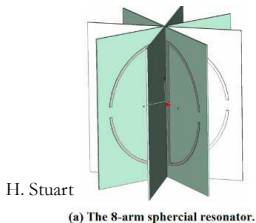
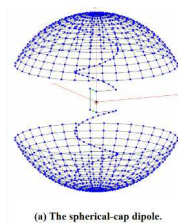
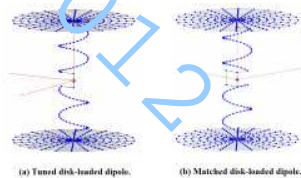
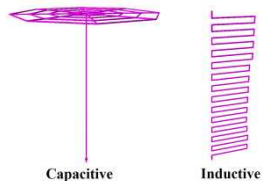
Lumped element

Distributed

Material Loading

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Tuning – Achieving Self Resonance



H. Stuart

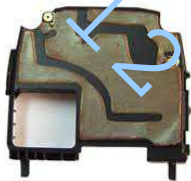
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Matching Techniques

- ❑ Shunt (parallel) stub or post (e.g. Inverted-F, PIFA)
- ❑ Elevated feed point
- ❑ Lumped element reactive matching at feed point
- ❑ Near-field reactive transformer
- ❑ Increasing turns (e.g. loop)
- ❑ Folding (e.g. dipole, monopole)

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Questions

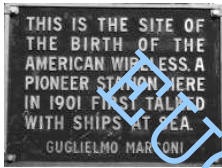


- ❑ After 100+ years, where do today's antenna engineers derive their design building blocks?
- ❑ What are some of today's engineering design challenges?

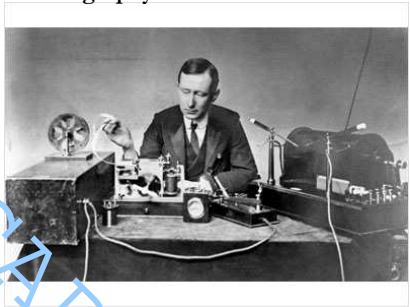
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Wireless Telegraphy

“To Whom it May Concern: Be it known that I, Guglielmo Marconi, a subject of the King of Italy, and a resident of London, England, have invented certain new and useful Improvements in Wireless Telegraphy.....”



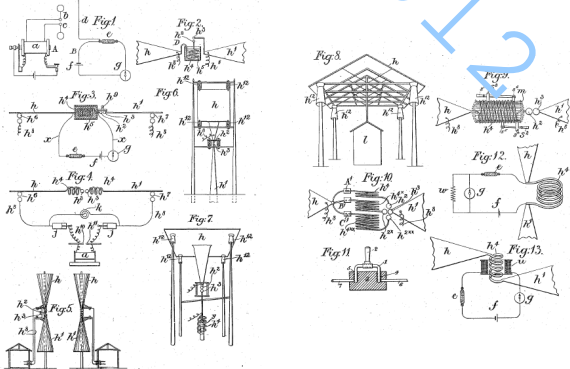
1901



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Wireless Telegraphy - 1898

Oliver Lodge's 609,154 patent (1898) – *Electric Telegraphy*



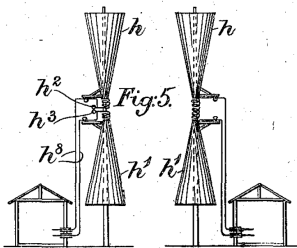
10

Wireless Telegraphy - 1898

Oliver Lodge's 609,154 patent (1898) – *Electric Telegraphy*

As charged surfaces or capacity areas spheres or square plates or any other metal surfaces may be employed; but I prefer, for the purpose of combining low resistance with great electrostatic capacity, cones or triangles or other such diverging surfaces with the vertices adjoining and their larger areas spreading out into space; or a single insulated surface may be used in conjunction with the earth, the earth or conductors embedded in the earth constituting the other oppositely-charged surface.

Radiation from an oscillator consisting of a pair of capacity areas is greater in the equatorial than in the axial direction, and accordingly the sending in all directions is desired it is well to arrange the axis of the emitter vertical. Moreover, radiation polarized in a horizontal plane—that is, with its electric oscillations vertical—is less likely to be absorbed during its passage over a partially-conducting earth or water. A pair of insulated capacity areas arranged for long-distance signaling is shown on the left-hand side of Fig. 5. Fig. 6 shows a single insu-



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Wireless Telegraphy - 1904

“According to my present invention I connect to the vertical elevated conductor which, as shown, is in the form of a wire or rod, one or more members consisting of wires or rods extending there from in a lateral or outward direction. The vertical portion of the aerial may then, if desired, be made comparatively short, and the necessary length and capacity secured by lateral extension.”

No. 760,463.

PATENTED MAY 24, 1904.

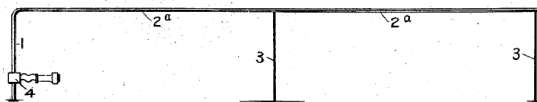
G. MARCONI.

WIRELESS SIGNALING SYSTEM.

APPLICATION FILED SEPT. 10, 1903.

NO MODEL.

FIG. 1.

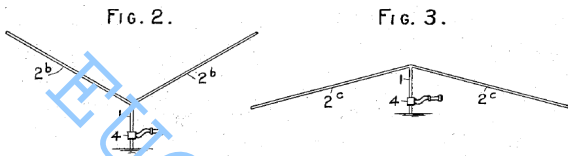


Marconi patented the Inverted-L antenna in 1904

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Wireless Telegraphy - 1904

“According to my present invention I connect to the vertical elevated conductor which, as shown, is in the form of a wire or rod, one or more members consisting of wires or rods extending there from in a lateral or outward direction. The vertical portion of the aerial may then, if desired, be made comparatively short, and the necessary length and capacity secured by lateral extension.”

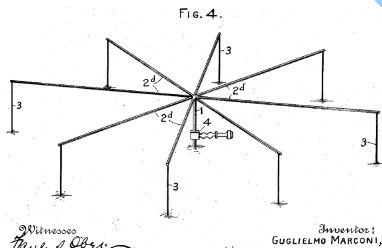


Marconi patented the Top-Loaded antenna in 1904

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Wireless Telegraphy - 1904

“According to my present invention I connect to the vertical elevated conductor which, as shown, is in the form of a wire or rod, one or more members consisting of wires or rods extending there from in a lateral or outward direction. The vertical portion of the aerial may then, if desired, be made comparatively short, and the necessary length and capacity secured by lateral extension.”



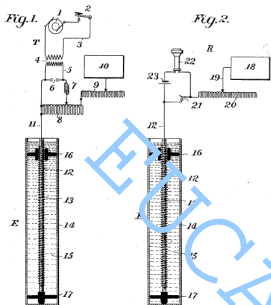
Marconi patented the Top-Loaded antenna in 1904

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Wireless Telegraphy - 1909

J. Murgas' 915,993 patent (1909) – *Wireless Telegraphy*

915,993. J. MURGAR. WIRELESS TELEGRAPHY. APPLICATION FILED MAY 13, 1907. Patented Mar. 23, 1909.

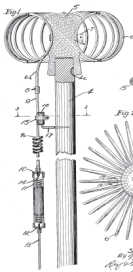


“My present improvement is directed to rendering the apparatus equally sensitive to waves of any length and any frequency and also to **reducing the length** or depth of the well or hole for the antenna and therefore its cost. I accomplish these objects by **coiling the antenna into a helix having an axis much shorter than the straight antenna** of the application mentioned.

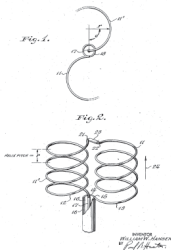
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Small Antennas

May 5, 1931. S. W. JESTER. HELIX. FIG. 1. Filed March 29, 1927. Sept. 27, 1943. W. W. HANSEN. SHORT STUBBY HELICAL COIL. FIG. 1. Filed Sept. 6, 1940. 2,482,797

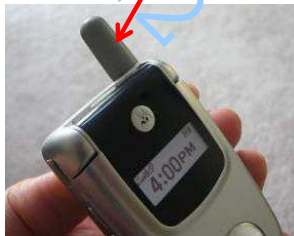


Filed 1927



Filed 1943

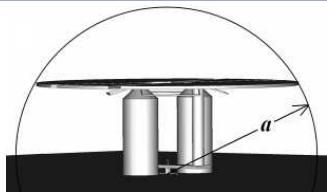
Short “stubby” helical coil



Helical Coil Antennas

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Small Antennas



$$k = \frac{2\pi}{\lambda}$$

a is the radius of a sphere circumscribing the maximum dimension of the antenna.

Electrically Small: $ka < 1$ (Wheeler – 1947); $ka < 0.5$ (Best – 2003)

$$Q_{lb} = \frac{1}{(ka)^2}$$

Wheeler – 1947

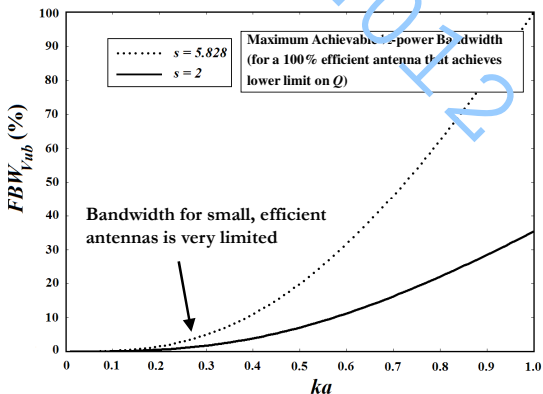
$$Q_{lb} = \frac{1}{(ka)^3} + \frac{1}{ka}$$

Chu – 1948; McLean - 1996

Fundamental Limits Defined for Small Antennas (1947-1948)

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Small Antennas



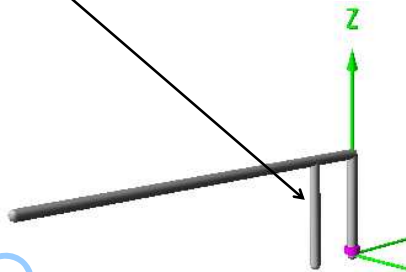
Upper Bound on Bandwidth

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Small Antennas

“Post” or “stub” to facilitate impedance matching of the Inverted-L

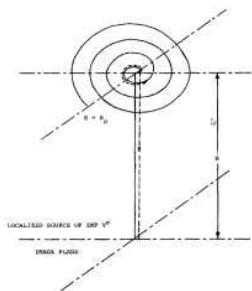
R. W. P King - 1960



Inverted-L Antenna

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Small Antennas



Bhojwani, 1973

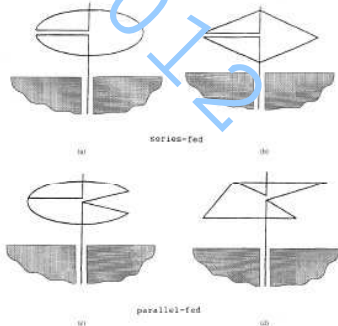


Fig. 1. Monopole loaded with key slots.

Altshuler, 1996

Inductive and Capacitive Loading

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Small Antennas

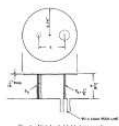


Fig. 7—Flat-backed helical antenna.

Seeley, 1956

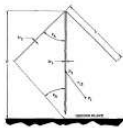


Fig. 8—Vertical helical antenna above a ground plane.

Gengi, 1965



Fig. 9—Chebyshev antenna for multibeam radiation.

Simpson, 1977

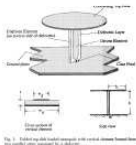


Fig. 10—Helical antenna loaded with a circular disc above a ground plane.

McLean, 1996

Inductive and Capacitive Loading

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Small Antennas

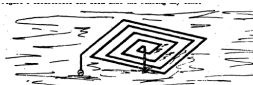


Figure 3. Electrically small antenna with simple flat spiral winding, an electrical half wave long

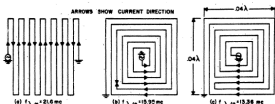


Figure 7. Three windings of same dimensions and containing the same length of wire, with their resonant frequencies

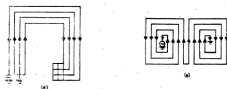


Figure 6. Windings in which mutual inductions between adjacent wires are positive

Fenwick, 1965



Fig. 8. (a) VSWR antenna wire model. (b) Schobert's antenna model. (c) No active impedance model.

Wanselow, 1966



Fig. 10—Loaded top-hat antenna loaded with a circular disc above a ground plane.

Harrison, 1963

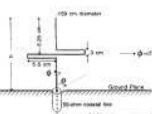


Fig. 1. Monopole antenna loaded with a modified helical dipole.

Altshuler, 1993

Inductive and Capacitive Loading

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Small Antennas

<p>1. SPHERICAL</p>	<p>The spherical loaded resonator is antenna has been tested by Kell and Pridmore (1970) and compared with a resonator of the same length as the antenna. The results are shown in the figure. The application of these resonators to HF antennas is discussed by Kell and Pridmore (1970). Further comments on antenna antennas and gain are given in the context of integrated antenna systems as described by Hansen (1981).</p>	<p>2. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>The HFA is used as HF and VLF having the advantage over many other integrated antennas, of that it is well suited and suitable for testing antennas. Hansen and Kelly (1970) give the radiation characteristics and losses. Theoretical efficiency values are given in the figure. The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>
<p>3. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>Large (1971) has calculated the reactance dependence of series-resonant type loaded resonators. These resonators have approximately the same radiation resistance as the resonator. It can be readily operated below self-resonance and consequently need external matching networks.</p>	<p>4. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>This VHF antenna is loaded with dielectric material to obtain a very low physical height and other means to reduce the radiation resistance. The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>
<p>5. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>This antenna is also based on the filament resonator but uses a filament of dielectric material. The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>	<p>6. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>The essential is to a special case of the transmission line antenna. The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>

Inductive and Capacitive Loading

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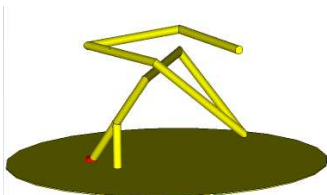
Small Antennas

<p>10. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>The loop may be resonant to inductively, the antenna is resonant to inductively. The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>	<p>11. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>
<p>12. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>	<p>13. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>
<p>14. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>	<p>15. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>
<p>16. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>	<p>17. SPHERICAL LOADED RESONATOR ANTENNA</p>	<p>The antenna is shown with a resonance of 10 MHz with a reactance of 8.5 ohms, which corresponds to 100 nH.</p>

Inductive and Capacitive Loading

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Recent Activities

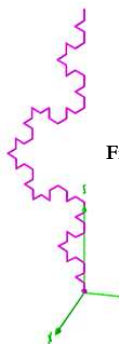


Optimization Algorithms:

Genetic Algorithms

Particle Swarm Optimization

Others



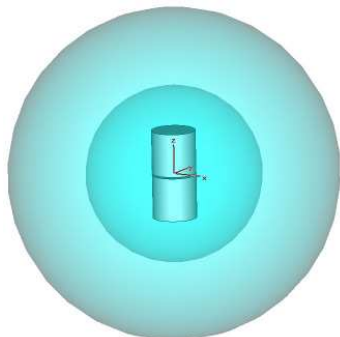
Fractal Antennas

Koch Fractal

Mid to Late 1990's: Geometric Considerations

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Recent Activities



Many definitions of
“Metamaterial” within the
community

No electrically small,
isotropic, homogeneous
material with negative
permittivity or permeability
exists

“Metamaterial-Inspired”

Mid 2000's – Present: Metamaterial Antennas

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Today's Building Blocks

Liu, Hall and Wake – 1997

Offers advantages of wider bandwidth and multiple operating bands.

Inverted-L (1904)

Inverted-F (1960 or before?)

PIFA (1980's or before?)

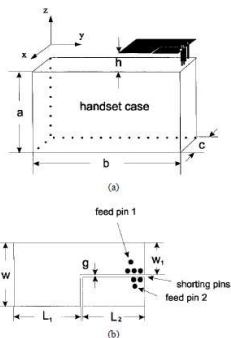


Fig. 1. (a) Geometrical arrangement of the dual-band antenna mounted on the conducting telephone case. (b) Top view of the dual-band antenna.

Planar Inverted-F Antenna (PIFA)

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Today's Small Antenna Perspective



While today's small antennas may be sophisticated, unique and exhibit near optimum performance, today's antenna engineers must move beyond the concept and design of the isolated antenna in free space or an isolated antenna on a ground plane. **Today's antenna engineers must understand and design the antenna as part of a much larger system.**

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Today's Wireless Communications



- Software Defined Radio
- Radio on a Chip
- Small Form Factor
- Wideband/Multiband
- Reconfigurable

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Today's Wireless Communications



Nokia 8390/8310
(Early Internal Antenna)

GSM Bands

GPS

UMTS

WiFi (802.11)

Bluetooth

LTE

FM Radio

System	Band	Uplink (MHz)	Downlink (MHz)	Channel number
T-GSM	380	380.2-389.8	350.2-399.8	dynamic
GSM-410	410	412-419.8	420.2-429.8	dynamic
GSM-450	450	450.4-457.6	460.4-467.6	259-293
GSM-480	480	478.8-486	488.8-496	300-340
GSM-710	710	698.0-710	728.0-740	dynamic
GSM-750	750	747.0-752.0	717.0-732.0	438-511
T-GSM-810	810	806.0-821.0	810-866.0	dynamic
GSM-850	850	840-849.0	869.0-894.0	128-251
P-GSM-900	900	890.0-915.8	935.2-959.8	1-124
E-GSM-900	900	880.0-914.8	925.2-959.8	975-1023, 0-124
R-GSM-900	900	878.0-914.8	921.0-959.8	955-1023, 0-124
T-GSM-900	900	870.4-876.0	915.4-921.0	dynamic
DCS-1800	1800	1710.2-1784.8	1805.2-1879.8	512-885
PCS-1900	1900	1850.0-1910.0	1930.0-1990.0	512-810

- Internal Device Dependent Designs
- Wideband/Multiband
- Small Form Factor
- User Dependent Effects
- MIMO
- Reconfigurable

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Today's Small Antenna Challenges



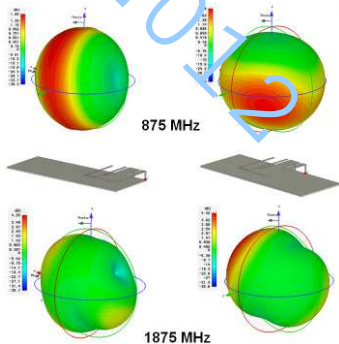
Today's wireless devices come in many different form factors. While meeting required performance properties, the antenna must be designed to fit within the electrical and mechanical constraints of the device.

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Today's Small Antenna Challenges



The front of the circuit board



Time-varying current within the device (ground plane, circuit board, etc) radiate and become part of the antenna system.

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Today's Small Antenna Challenges

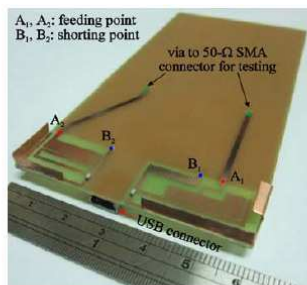
Introduction of iPhone 4



Performance and design of the wireless system must include the affects of the user.

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Today's Small Antenna Challenges



"Internal Handset Antenna Array for LTE/WWAN and LTE MIMO Operations,"
Ting-Wei Kang*, Kin-Lu Wong, Ming-Fang Tu,
EuCap 2011

Fig. 2 Photo of the fabricated antenna array (handset casing not included in the photo).

Antenna engineers must do more than just design an "antenna". Oftentimes, the antenna engineer must focus on optimizing system performance.

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Today's Small Antenna Designs

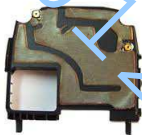


100+ years of fundamental design approaches allows today's antenna engineers to develop sophisticated antennas as part of designing and optimizing a complete wireless system

- Design antennas within the mechanical constraints of the device
- Design the antenna to operate at multiple communication bands
- Optimize the antenna when considering user dependent effects
- Understand the propagation environment and channel
- Understand that the device is part of the radiating structure
- Optimize system performance/capacity using reconfigurable antennas, MIMO antennas, etc.

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Today's Small Antenna Designs



What's Next/Current Activities?

- High permeability, low loss materials
- Non-Foster Impedance matching
- Small multi-resonant/wideband antennas
- Smart, reconfigurable small antenna – tunes and matches itself depending on waveform bandwidth, operating frequency, changes in surrounding environment, etc.

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Today's Small Antenna Designs



What is certain: Not considering the wireless device as a complex, integrated system operating in a complex environment - an optimum system design may not be achieved